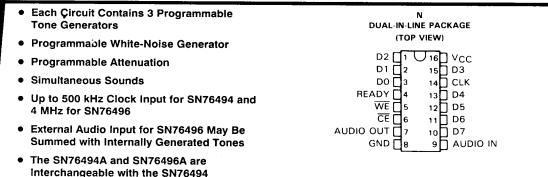
SN76494, SN76494A, SN76496, SN76496A PROGRAMMABLE TONE/NOISE GENERATOR

D2801, JUNE 1984 - REVISED JANUARY 1989



description

and SN76496, Respectively

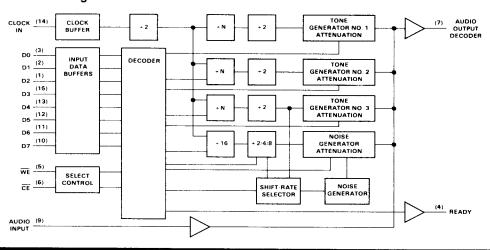
The SN76494 and SN76494A digital complex sound generators are integrated injection logic (I²L) tone generators designed to provide low-cost tone or noise generation capability in microprocessor systems. The SN76494 and SN76494A are data-bus-based input-output peripheral devices that interface the microprocessor through 8 data lines and 3 control lines.

The SN76494 and SN76494A are identical to the SN76496 and SN76496A except that the maximum clock input frequency for SN76494 and SN76494A is 500 kHz and for SN76496 and SN76496A, it is 4 MHz. A "divide-by-eight" stage is deleted from the SN76496 and SN76496A circuitry so that only 4 clock pulses are required to load the data into the SN76494 and SN76494A, compared to 32 pulses for the SN76496 and SN76496A.

Either of these devices may also be used as a replacement for the SN76489A in all applications if pin 9 is left open or grounded. The output load must be limited to 10 mA.

When audio input is not desired in the SN76494, SN76494A, SN76496 or SN76496A, the audio input pin should be grounded.

functional block diagram



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TEXAS INSTRUMENTS

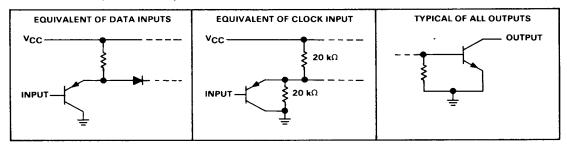
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SN76494, SN76494A, SN76496, SN76496A PROGRAMMABLE TONE/NOISE GENERATOR

schematics of inputs and outputs



absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

| Supply voltage, VCC (see Note 1) 7 V |
|---|
| Input voltage, V _I : Audio input |
| All other inputs |
| Output current at pin 7 |
| Continuous total power dissipation at (or below) 25°C free-air temperature (see Note 2) 1150 mW |
| Operating free-air temperature range |
| Storage temperature range |
| Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds |

NOTES: 1. All voltage values are with respect to network ground terminal.

recommended operating conditions

| | | SN764 | SN76494, SN76494A | | | 96, SN7 | 6496A | |
|-----------------|------------------------------------|-------|-------------------|-----|-----|---------|-------|----------|
| | | MIN | NOM | MAX | MIN | NOM | MAX | UNIT |
| Vcc | Supply voltage | 4.5 | 5 | 5.5 | 4.5 | 5 | 5.5 | V |
| VIH | High-level input voltage | 2 | | | 2 | | | ٧ |
| VIL | Low-level input voltage | | | 0.8 | | | 0.8 | |
| lj. | Audio input current | 0 | | 1.8 | 0 | | 1.8 | mA |
| VOH | High-level output voltage (pin 4) | | | 5.5 | | | 5.5 | V |
| lOL | Low-level output current (pin 4) | | • | 2 | | | 2 | mA |
| fclock | Input clock frequency | | | 0.5 | | | 4 | MHz |
| td(WE) | Delay time, CE low to WE low | 0 | | | 0 | | | ns |
| t _{su} | Setup time, data before WE↓ or CE↓ | 0 | | | 0 | | | ns |
| th | Hold time, data after READY† | 0 | | | 0 | | | ns |
| TA | Operating free-air temperature | 0 | | 70 | 0 | | 70 | °C |



^{2.} For operation above 25°C free-air temperature, derate linearly to 736 mW at 70°C at the rate of 9.2 mW/°C.

electrical characteristics over recommended ranges of supply voltage and operating free-air temperature range (unless otherwise noted)

| | PARAM | METER | nput V ₁ = 0 -25 -175 | | | | | |
|------------------|---|--------------------------|---|-----|------|------|----|--|
| ЮН | High-level output current (p | oin 4) | V _O = 5.5 V | - | | 10 | μΑ | |
| ۱н | High-level input current (Al | digital inputs) | $V_I = V_{CC}$ | | | 10 | μA | |
| tes | Low-level input current | ČE input | | | -25 | -175 | | |
| ll L | LOW-level hiput current | All other digital inputs | V _I = 0 | | -10 | -70 | μΑ | |
| V_{IB} | Input bias voltage, audio (p | oin 9) | $R = 4.7 \text{ k}\Omega \text{ to V}_{CC}$ | 5. | | | | |
| Vон | High-level output voltage (p | oin 7) | | | | 5.5 | V | |
| VOL | Low-level output voltage (p | in 4) | I _{OL} = 2 mA | | 0.25 | 0.4 | V | |
| V _{OPP} | Peak-to-peak output voltag | e (pin 7) | Attenuation: | 260 | | | mV | |
| lcc | Low-level output voltage (pin 4) Peak-to-peak output voltage (pin 7) Supply current 2 dB NOM 4 dB NOM | | | | 30 | 50 | mA | |
| | | 2 dB NOM | | 1 | 2 | 3 | | |
| | Attenuation | 4 dB NOM | Con Table 4 | 3 | 4 | 5 | | |
| | Attenuation | 8 dB NOM | See Table 1 | 7 | 8 | 9 | dB | |
| | | 16 dB NOM | | 15 | 16 | 17 | . | |
| Ci | Input capacitance | | | | | 15 | pF | |

[†] Typical values are at V_{CC} = 5 V, T_A = 25°C.

switching characteristics, VCC = 5 V, TA = 25°C

| | PARAMETER | TEST CONDITIONS | MIN | TYP | MAX | UNIT |
|------------------|---|--|-----|-----|-----|------|
| tPHL | Propagation delay time, high-to-low level RDY output from CE | | | 90 | 150 | ns |
| ^t PHL | Propagation delay time high-to-low level, RDY output from WE | C_L = 225 pF, R_L = 2 k Ω to V_{CC} | | 90 | | ns |
| ^t PLH | Propagation delay time low-to-high level, RDY output from CLK | | | 90 | | ns |

PARAMETER MEASUREMENT INFORMATION

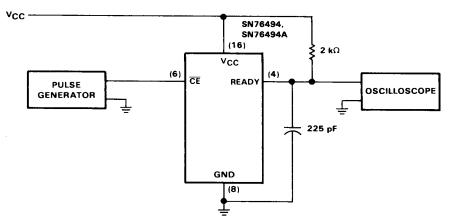


FIGURE 1. tPHL TEST CIRCUIT



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pin assignments and functions

| SIGNATURE | PIN | 1/0 | DESCRIPTION |
|-----------|-----|-----|---|
| CE | 6 | 1 | Chip Enable. When chip enable is low, the device is operational, input terminals are enabled, and data may be entered. |
| D0 (MSB) | 3 | ı | |
| D1 | 2 | 1 | |
| D2 | 1 | | |
| D3 | 15 | 1 | Do shough D7. I have dista hun |
| D4 | 13 | 1 | D0 through D7 — Input data bus |
| D5 | 12 | 1 | |
| D6 | 11 | 1 | |
| D7 (LSB) | 10 | ı | |
| Vcc | 16 | 1 | Supply voltage (5 V nom) |
| GND | 8 | 0 | Ground reference |
| CLOCK | 14 | ı | Input Clock |
| WE | 5 | ı | Write Enable. When \overline{CE} is enabled and \overline{WE} is active (low), input on the data bus is accepted. \overline{CE} and \overline{WE} must be held low until READY returns high (four clock cycles for the SN76494 and SN76494A or 32 clock cycles for the SN76496 and SN76496A). If \overline{WE} remains low throughout four additional clock cycles for the SN76494 and SN76494A (32 clock cycles for the SN76496 and SN76496A) a new write cycle will be initiated. |
| READY | 4 | 0 | When low, READY indicates that a write cycle is in progress; data on the input bus must remain valid until READY returns high. |
| AUDIO IN | 9 | i | Audio input from external source |
| AUDIO OUT | 7 | 0 | Audio Drive Out |

PRINCIPLES OF OPERATION

tone generators

Each tone generator consists of a frequency synthesis section and an attenuation section. The frequency synthesis section requires 10 bits of information (F0-F9) to define half the period of the desired frequency (f). F0 is the most significant bit and F9 is the least significant bit. This information is loaded into a 10-stage tone counter, which counts down at an N/2 rate where N is the input clock frequency. When the tone counter counts down to zero, a borrow signal is produced. This borrow signal toggles the frequency flip-flop and also reloads the tone counter. Thus, the period of the desired frequency is twice the value of the period register.

The frequency can be calculated by the following:

$$f = \frac{N}{4n}$$
 for SN76494 and SN76494A, or $f = \frac{N}{32n}$ for SN76496 and SN76496A

where N = clock in Hz n = 10-bit binary number

The output level of each tone/noise generator may be selected by programming a four stage attenuator. The attenuator values, along with their bit position in the data word, are shown in Table 1. Multiple attenuation control bits may be true simultaneously. Thus, the maximum attenuation is 30 dB.



TABLE 1. ATTENUATION CONTROL

| | BIT PO | WEIGHT | | |
|----|--------|--------|----|---------|
| A0 | A1 | A2 | А3 | (in dB) |
| 0 | 0 | 0 | 1 | 2 |
| 0 | 0 | 1 | 0 | 4 |
| 0 | 1 | 0 | 0 | 8 |
| 1 | 0 | 0 | 0 | 16 |
| 1 | 1 | 1 | 1 | OFF |

noise generator

The noise generator consists of a noise source and an attenuator. The noise source is a shift register with an exclusive OR-feedback network. The feedback network has provisions to protect the shift register from being locked in the zero state.

TABLE 2. NOISE FEEDBACK CONTROL

| FEEDBACK | CONFIGURATION |
|----------|------------------|
| 0 | "Periodic" noise |
| 1 | "White" noise |

Whenever the noise control register is changed, the shift register is cleared. The shift register will shift at one of four rates as determined by the two NF bits. The fixed shift rates are derived from the input clock.

TABLE 3. NOISE GENERATOR FREQUENCY CONTROL

| BI | TS | SHIFT RATE |
|-----|-----|--------------------------|
| NFO | NF1 | SHIFTRATE |
| 0 | 0 | N/64 |
| 0 | 1 | N/128 |
| 1 | 0 | N/256 |
| 1 | 1 | Tone generator #3 output |

The output of the noise source is connected to a programmable attenuator as shown in Figure 4.

output buffer/amplifier

The output buffer is a conventional operational amplifier summing circuit. It sums the three tone generator outputs, the noise generator output, and any audio input through pin 9. The output buffer will generate up to 10 mA.

To prevent oscillations in the output buffer, the output (pin 7) should be decoupled. This is done by putting 10 ohms in series with 0.1 µF from pin 7 to ground (see Figure 3).

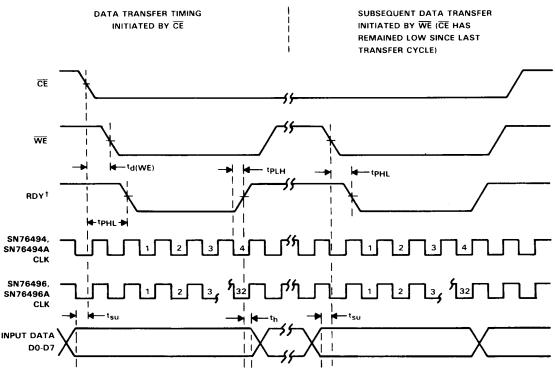
data transfer

The microprocessor selects the SN76494, SN76494A, SN76496, or SN76496A by taking \overline{CE} low (low voltage). Unless \overline{CE} is low, no data transfer can occur. When \overline{CE} is low, the \overline{WE} signal strobes the contents of the data bus to the appropriate control register. The data bus contents must be valid at this time.



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The SN76494 and SN76494A require approximately 4 clock cycles to load the data into the control register. The SN76496 and SN76496A require approximately 32 clock cycles. The open-collector READY output is used to synchronize the microprocessor to this transfer and is pulled to the false state (low) immediately following the falling edge of $\overline{\text{CE}}$ (or $\overline{\text{WE}}$ when data transfer is initiated by $\overline{\text{WE}}$). READY will go high upon completion of the data transfer cycle. The data transfer timing is shown below.



[†] WE must be returned high (inactive) within 4 clock pulses for the SN76494 and SN76494A (32 clock pulses for the SN76496A) after RDY returns high. Otherwise, a new data transfer cycle will be initiated.

FIGURE 2. DATA TRANSFER TIMING

TABLE 4. FUNCTION TABLE

| INP | UTS | OUTPUT | This table is valid when the | | | | |
|-----|----------|--------|----------------------------------|--|--|--|--|
| CE | WE READY | | device is: | | | | |
| Ĺ | L | L | (1) not being clocked, and | | | | |
| L | Н | L | (2) is initialized by pulling WE | | | | |
| Н | L | н | and CE high. | | | | |
| н | н | H | | | | | |



CPU interface to SN76494, SN76494A, SN76496 or SN76496A

The microprocessor interfaces with the SN76494, SN76494A, SN76496A by means of the 8 data lines and 3 control lines (WE, CE and READY). Each tone generator requires 10 bits of information to select the frequency and four bits of information to select the attenuation. A frequency selection requires a double-byte transfer, while an attenuator selection requires a single-byte transfer.

If no other control registers on the chip are accessed, a tone generator may be rapidly updated by initially sending both bytes of frequency and register data, followed by just the second byte of data for succeeding values. The register address is latched on the chip, so the data will continue going into the same register. This allows the six most significant bits to be quickly modified for frequency sweeps.

control registers

The devices have 8 internal registers that are used to control the 3 tone generators and the noise source. During all data transfers to the devices, the first byte contains a 3-bit field that determines the destination control register. The register address codes are shown in Table 5.

TABLE 5. REGISTER ADDRESS FIELD

| R0 | R1 | R2 | DESTINATION CONTROL REGISTER |
|----|----|----|---------------------------------|
| 0 | 0 | 0 | Tone 1 Frequency |
| 0 | 0 | 1 | Tone 1 Attenuation |
| 0 | 1 | 0 | Tone 2 Frequency |
| 0 | 1 | 1 | Tone 2 Attenuation |
| 1 | 0 | 0 | Tone 3 Frequency |
| 1 | 0 | 1 | Tone 3 Attenuation |
| 1 | 1 | 0 | Noise Control |
| 1 | 1 | 1 | Noise Attenuation |

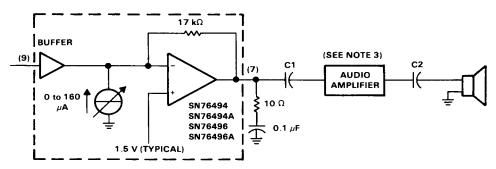
data formats

The formats required to transfer data are shown below.

| | REG ADDR DAT | | | TA | | | | | | DATA | | | | | |
|-------|---------------------|---------------|---------|--------|--------|--------|-------------|-------|---|------|-----|-------|------|----|-------|
| 1 | Ro | R1 | R2 | F6 | F7 | F8 | F9 | 0 | x | F0 | F1 | F2 | F3 | F4 | F5 |
| BIT 0 | | FIRST | BYTE | | | BIT 7 | | BIT 0 | | | SEC | OND E | SYTE | | BIT 7 |
| UPDA' | UPDATE NOISE SOURCE | | | | (SING | LE BYT | E TRANSFER) | | | | | | | | |
| | REG ADDR | | | | SH | HFT | | | | | | | | | |
| 1 | R0 | R1 | R2 | × | FB | NF0 | NF1 | | | | | | | | |
| BIT 0 | • | | | • | | | BIT 7 | | | | | | | | |
| UPD | ATE AT | TENUA | ATOR (S | SINGLE | BYTE ' | TRANS | FER) | | | | | | | | |
| | R | REG ADDR DATA | | | | | | | | | | | | | |
| 1 | RO | R1 | R2 | A0 | A1 | A2 | A3 | | | | | | | | |
| BIT 0 | | | | | | | BIT 7 | | | | | | | | |



TYPICAL APPLICATION DATA



NOTE 3: The capacitance values of C1 and C2 are determined by the frequency response desired and the audio amplifier used.

CLOCK -**Vcc** SN76494, SN76494A. DO SN76496, Vсс READY OR SN76496A WE 2.5 kΩ ₹ CÉ SELECT -AUDIO OUT AUDIO IN-D6

FIGURE 3. EXTERNAL AUDIO OUTPUT INTERFACE

- NOTES: 4. The data lines must be latched so that the data remains on them at least 32 clock cycles for the SN76496 and SN76496A or (4 clock cycles for the SN76494 and SN76494A) after the select line goes low.
 - 5. The select pulse should be a negative-going pulse with minimum duration of 150 ns.

FIGURE 4. MICROCOMPUTER PARALLEL PORT INTERFACE



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